# Kalman Filtering and Riccati Equations for Descriptor Systems

Ramine Nikoukhah, Member, IEEE, Alan S. Willsky, Fellow, IEEE, and Bernard C. Levy, Senior Member, IEEE

Abstract—In this paper, we consider a general formulation of a discrete-time filtering problem for descriptor systems. It is shown that the nature of descriptor systems leads directly to the need to examine singular estimation problems. Using a "dual approach" to estimation we derive a so-called "3-block" form for the optimal filter and a corresponding 3-block Riccati equation for a general class of time-varying descriptor models which need not represent a well-posed system in that the dynamics may be either over or under constrained. Specializing to the time-invariant case we examine the asymptotic properties of the 3-block filter, and in particular analyze in detail the resulting 3-block algebraic Riccati equation, generalizing significantly the results in [23], [28], [33]. Finally, the noncausal nature of discrete-time descriptor dynamics implies that future dynamics may provide some information about the present state. We present a modified form for the descriptor Kalman filter that takes this information into account.

#### I. INTRODUCTION

In this paper, we address the problem of recursive estimation for a general class of descriptor systems. Specifically the systems that we consider are of the form<sup>1</sup>

$$E_{k+1}x(k+1) = A_k x(k) + u(k), \quad k \ge 0$$
 (1.1)

$$y(k+1) = C_{k+1}x(k+1) + r(k), \qquad k \ge 0 \quad (1.2)$$

where the matrix  $E_{k+1}$  is  $l_k \times n_{k+1}$ ,  $A_k$  is  $l_k \times n_k$ , and  $C_{k+1}$  is  $p_{k+1} \times n_{k+1}$ . Here *u* and *r* are zero-mean white Gaussian noise sequences with

$$\mathscr{M}\left[\binom{u(k)}{r(k)}\binom{u(j)}{r(j)}^{T}\right] = \binom{Q_{k} \quad S_{k}}{S_{k}^{T} \quad R_{k}}\delta(k-j) \quad (1.3)$$

Manuscript received February 15, 1991; revised January 29, 1992. Paper recommended by Past Associate Editor, B. K. Ghosh. This work was supported in part by the Air Force Office of Scientific Research under Grant AFOSR-88-0032, the Army Research Office under Grant DAAL-03-86-K0171, the National Science Foundation under Grants ECS-8700903 and ECS-8700203, and the Office of Naval Research under Grant USN-N00014-89-J-3135.

R. Nikoukhah is with INRIA, Domaine de Voluceau, Rocquencourt, BP 105, 78153 Le Chesnay Cedex, France.

 A. S. Willsky is with the Department of Electrical Engineering and Computer Science and the Laboratory for Information Decision Systems, Massachusetts Institute of Technology, Cambridge, MA 02139.
 B. C. Levy is with the Department of Electrical Engineering and

B. C. Levy is with the Department of Electrical Engineering and Computer Science, University of California, Davis, CA 95616.

IEEE Log Number 9202203.

<sup>1</sup> The indexing choices in (1.1) and (1.2) have been made in part to simplify the subsequent development. For example, the use of the notation of r(k) in (1.2) rather than r(k + 1) is consistent with and simplifies (1.3). Specifically, with these choices, the noises u(k) and r(k) are possibly correlated, and both affect the information we acquire on x(k + 1).

where  $\mathcal{M}(\cdot)$  denotes the mean and  $\delta(k) = 1$  if k = 0 and 0 otherwise. Also, we assume that x(0) is a Gaussian random variable independent of  $(u^T(k), r^T(k))^T$ , with mean  $\bar{x}_0$  and covariance  $P_0$ , independent of  $(u^T(k), r^T(k))^T$ .<sup>2</sup> The problem that we consider in this paper is the recursive estimation of x(k).

Several aspects of this model deserve a comment. The study of descriptor systems, of course, has a rich and growing literature [9], [10], [18]. Some of the motivation for this activity comes from applications in which the natural descriptions of systems involve both dynamics and constraints among variables, leading to models of the form (1.1) with a possibly singular matrix E on the lefthand side. Furthermore, in studies such as [35] it is argued that models of this type are a natural starting point for modeling when we are attempting to deduce relations among dynamically evolving quantities rather than imposing causative structure. Indeed in [19]-[21] as well as in our previous work [23]-[29], it has been emphasized that descriptor models such as (1.1) can be used to describe noncausal phenomena, e.g., where the variable "k" represents space rather than time which typically involve such dynamics together with boundary conditions. In fact, in a subsequent paper [26], the results that we developed here are used for constructing efficient smoothing algorithms for such boundary-value models.

A second point to note concerning the model (1.1) is that it allows the possibility that the dimensions of the problem [the number  $l_k$  of dynamic constraints, the number  $p_k$  of measurements, and the dimension  $n_k$  of x(k)] may vary with k. As we shall see, this does not cause any difficulty in our analysis, but this is not our reason for including this level of generality. A better reason is that such a situation arises naturally in "recursive" descriptions of two-dimensional (2-D) phenomena. Specifically, as shown in [16], there are very natural directions of recursion for boundary-value models, namely in from, or out towards the boundary. The inward propagation, for example, involves propagating boundary conditions into the domain of interest. In 1-D problems, where the boundary consists of two points, inward propagation leads to a new boundary which also has two points. In 2-D

0018-9286/92\$03.00 © 1992 IEEE

<sup>&</sup>lt;sup>2</sup> The *a priori* information on the initial condition can also be modeled as an additional observation of the form (1.2), see Section III for more details. Final and intermediate conditions (if any) can also be incorporated into observations (1.2).

however, the boundary of a compact domain *changes size* as we shrink or expand the domain. Thus, if we think of x(k) as representing the values of a process along such a shrinking or expanding boundary, we see that we have no choice but to deal with changing dimensionality. While we do not explicitly focus on such problems in this paper, our analysis contains all the elements necessary to make it directly applicable to 2-D estimation problems.

Another reason for allowing the possibility of changing dimensionality is one that has other even more important implications. Specifically, rather than thinking of (1.1) as describing system dynamics, we may wish to think of it as providing a set of possibly noisy constraints on the behavior of x. From this perspective, the information in (1.1)plays essentially the same role as that in (1.2), the only apparent difference being that each piece of information in (1.1) concerns x at two consecutive points in time rather than the single-point-in-time nature of the information (1.2). From this perspective, allowing a change in dimensionality corresponds to allowing the possibility that at some points in time we might have more pieces of information than at others. Also, this perspective opens the question of the order in which these pieces of information are incorporated. For the most part, in this paper we will use the obvious ordering, namely (1.1) through time k to estimate x(k). However, there are other possibilities. In particular, in this paper we also consider the use of (1.1) over the entire time interval of interest, together with (1.2) through time k to estimate x(k). As we will see, when  $E_k = I$ , there is no difference between these two cases, but there is a difference when one considers the case of  $E_k$  singular, again emphasizing the noncausality of such models. Furthermore, the possible singularity of  $E_k$  coupled with the interpretation of (1.1) as an additional source of "measurement" data leads directly to the need to consider the possibility that some "measurements" are perfect. Thus, in our formulation we allow the possibility that  $R_k$  in (1.3) is singular.

Recursive estimation for descriptor systems has been the subject of several studies in recent years [8], [13], [23], [24], [28], [33]. In particular, in [24] we addressed this problem in the context of optimal smoothing for wellposed, constant coefficient boundary-value descriptor systems, i.e., systems of the form (1.1) and (1.2) which are square and constant (i.e.,  $l_k = n_k \equiv n, p_k \equiv p$ , and all matrices are constant), together with both the assumption that  $\{E, A\}$  is a regular pencil and a set of boundary conditions which yield a unique solution to (1.1) for any input u(k). In that paper, we used the method in [1] to derive a  $2n \times 2n$  Hamiltonian (boundary-value descriptor) system for the optimal smoother also assuming that the measurement noise covariance R was nonsingular. In addition, we introduced a new generalized algebraic Riccati equation and showed that if a solution to this equation existed, the Hamiltonian dynamics could be decoupled leading to parallel forward and backward recursions reminiscent of the Mayne-Fraser smoothing algorithm. In subsequent work in developing a system theory for such

systems, we obtained a set of necessary and sufficient conditions for the existence and uniqueness of positive definite solutions for this class of generalized Riccati equations [23], [28] and also provided a statistical interpretation for this solution. More recently, Wang and Bernhard [33] have developed some closely related results by dualizing their work on optimal control for descriptor systems [7]. Because of this perspective, less attention was paid to statistical interpretations of the results, and also their approach deals with estimating Ex(k) rather than x(k). On the other hand, Wang and Bernhard consider the more general case in which the pencil  $\{E, A\}$  need not be regular and in fact may not even be square (so that  $l \neq n$ ) and in this context develop analogous results on filter convergence and Riccati equations to those in [23], [28]

While the restriction to the estimation of Ex(k) is not significant if E is invertible, it is substantive if E is singular. Furthermore, as we have hinted, the possible singularity of E and A, together with the objective of estimating all of x(k), leads directly to the need to consider the possibility of perfect "measurements" either through the dynamics (1.1) or the observations (1.2). In this paper, we develop a procedure for optimal recursive estimation that is valid in the most general framework with E and A not necessarily square nor invertible and with possibly singular measurement noise covariances. As we will see, considering such a problem leads to the introduction of what we refer to as "3-block" forms for Hamiltonians, filters, and Riccati equations. Such forms actually can be found in various contexts in a number of papers in estimation and control [3]-[5], [14], [22], [32], [34]. Our work builds most directly on the approach of [34, chapter 11] and the machinery for singular estimation in [11] to derive not only a new 3-block Hamiltonian form valid in our general context but also a new 3-block generalized Riccati equation. In addition, in the constant dimension/constant matrix case, we develop convergence and steady-state results for the algebraic version of this equation, thereby extending the earlier results in [23], [28], and [33].

In the next section, we present and review some of the basic concepts concerning maximum likelihood parameter estimation with particular emphasis on deriving a form that is valid when some of the measurements are perfect. In Section III these results allow us to address the filtering problem for the system (1.1) and (1.2) resulting in the 3-block form for the descriptor Kalman filter and a corresponding 3-block Riccati equation. In Sections IV and V, we then focus on the time-invariant case. In the first of these sections, we generalize the results in [23], [28], and [33] by studying in detail the asymptotic properties of the descriptor Kalman filter. In particular we provide conditions for filter stability and for the convergence of the solution to the Riccati equation. Conditions under which the resulting 3-block algebraic Riccati equation has a unique positive semidefinite solution are given, and in Section V we generalize the well-known eigenvector approach to solving standard Riccati equations [31], [32] to our 3-block form. Finally, in Section VI, we show how the estimation procedure we have developed can be modified to account for the information about x(k) contained in the future dynamic constraints.

#### II. A LOOK AT MAXIMUM LIKELIHOOD ESTIMATION

In this section, we examine a few features of maximum likelihood (ML) linear estimation beginning with the simple problem of estimating an unknown n-dimensional vector x based on the p-dimensional measurement vector.

$$y = Hx + v \tag{2.1}$$

where v is a zero-mean Gaussian random vector with covariance R. While this is a well-studied problem it is worth making a few comments about it. First, note that the study of this estimation problem actually includes least-squares Bayesian estimation for Gaussian vectors. Specifically, consider the problem of computing the leastsquares estimate of a Gaussian random vector x with mean m and covariance P based on the measurement vector z = Cx + n where n is zero-mean Gaussian, independent of x, with covariance N. It is straightforward to check that this problem yields the same estimate as the ML problem with

$$y = \begin{pmatrix} m \\ z \end{pmatrix}, H = \begin{pmatrix} I \\ C \end{pmatrix}, R = \begin{pmatrix} P & 0 \\ 0 & N \end{pmatrix}.$$
 (2.2)

We focus here on the ML viewpoint, which in the next section will lead to our interpreting dynamic constraints as in (1.1) as additional pieces of information or measurements. A second point is that we focus here, for the most part, on the case in which x is *estimable*, i.e., in which (2.1) provides sufficient constraints so that we can in fact estimate all components of x. This is equivalent to assuming that H has rank  $n = \dim(x)$ , which, for example, is aways true in the Bayesian case (i.e., H in (2.2) obviously has rank n).

The third and most important point for us is that we wish to consider ML problems where R may not be of full rank. If H and R have full rank, the solution to the ML problem is easy to write out explicitly

$$\hat{x}_{\rm ML} = (H^T R^{-1} H)^{-1} H^T R^{-1} y.$$
 (2.3)

The error variance associated with this estimate is given by

$$P_{\rm ML} = \mathscr{M} \Big[ (x - \hat{x}_{\rm ML}) (x - \hat{x}_{\rm ML})^T \Big] = (H^T R^{-1} H)^{-1}.$$
(2.4)

The fact that the calculations can be described in such explicit form is extremely important as it allows us to obtain an explicit recursive structure for sequential estimation problems. What we would like to do is to obtain an equally explicit form when R is singular. To do this, we begin by recasting the ML estimation problem as a quadratic minimization problem. This approach, described

Let V be a full rank square root of R (so that  $VV^T = R$ ). Then we can write v = Vw, where w is a zero-mean Gaussian random vector with covariance I. If we then view the measurement

$$y = Hx + Vw \tag{2.5}$$

as a *linear constraint* on x and w, the ML problem is simply one of finding a pair (x, w) satisfying (2.5) and maximizing the probability density of w or, equivalently, minimizing

$$J(w) = (1/2)w^T w.$$
 (2.6)

This problem is readily solved using the Lagrange multipliers. Specifically, let

$$L(w, x, \lambda) = (1/2)w^{T}w + \lambda^{T}(y - Hx - Vw). \quad (2.7)$$

Setting the partials with respect to w, x, and  $\lambda$  to zero yields

$$w - V^T \lambda = 0 \tag{2.8a}$$

$$H^T \lambda = 0 \tag{2.8b}$$

$$-Hx - Vw = 0. \tag{2.8c}$$

Using (2.8a) to eliminate w gives the 2-block (p + n)-dimensional set of equations

v

$$\begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix} \begin{pmatrix} \lambda \\ x \end{pmatrix} = \begin{pmatrix} y \\ 0 \end{pmatrix}.$$
 (2.9)

A first obvious question about this set of equations concerns the invertibility of the  $(p + n) \times (p + n)$  matrix in (2.9). Note that one obvious necessary condition is that H must have full-column rank, as otherwise the last ncolumns would not be linearly independent. a second immediate necessary condition is that the first p rows must be linearly independent. The following shows that this pair of conditions is also sufficient.

Lemma 2.1: Let R be positive semidefinite and H a full-column rank matrix. Then, if  $[R \ H]$  has full-row rank, the matrix

$$\begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}$$
(2.10)

is invertible.

Proof: Suppose that

$$\begin{pmatrix} x^T & y^T \end{pmatrix} \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix} = 0.$$
 (2.11)

Then

and

$$x^{T}R + y^{T}H^{T} = 0 (2.12)$$

$$x^T H = 0. \tag{2.13}$$

If we now take the transpose of (2.13) and multiply it on the left by  $y^T$  we get

$$y^T H^T x = 0 \tag{2.14}$$

which after substitution in (2.12) postmultiplied by x yields is given by

$$x^T R x = 0 \tag{2.15}$$

which, since R is positive semidefinite gives  $x^{T}R = 0$ . Together with (2.13), this yields  $x^{T}[R \ H] = 0$ , which implies x = 0, since  $\begin{bmatrix} R & H \end{bmatrix}$  has full-row rank. Then (2.12) implies that  $y^T H^T = 0$ , and since H has full rank we have y = 0, so that (2.10) is invertible.

Assuming that (2.10) is invertible, we have from (2.9) that

$$\hat{x}_{\mathrm{ML}} = \begin{pmatrix} 0 & I \end{pmatrix} \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{-1} \begin{pmatrix} I \\ 0 \end{pmatrix} y \qquad (2.16)$$

and by direct calculation we find that  $\hat{x}_{\rm ML}$  is unbiased and has for error covariance

$$P_{\rm ML} = (0 \quad I) \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{-1} \begin{pmatrix} R & 0 \\ 0 & 0 \end{pmatrix} \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{-1} \begin{pmatrix} 0 \\ I \end{pmatrix}.$$
(2.17)

Writing

$$\begin{pmatrix} R & 0 \\ 0 & 0 \end{pmatrix} = \begin{pmatrix} R & 0 \\ H^T & 0 \end{pmatrix} - \begin{pmatrix} 0 & 0 \\ H^T & 0 \end{pmatrix}$$
(2.18)

we obtain the following simpler expression for  $P_{\rm ML}$ :

$$P_{\rm ML} = -(0 \quad I) \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{-1} \begin{pmatrix} 0 \\ I \end{pmatrix}.$$
(2.19)

The condition that  $[R \ H]$  has full-row rank has a simple physical interpretation, namely that there are no redundant perfect measurements (i.e., that no independent linear combinations of the observations yield noisefree measurements of the same linear combination of components of x). While this would certainly seem to be a reasonable assumption and can in principle be enforced by identifying redundancies and eliminating them, it is convenient to have a result that applies even when (2.10)is not invertible.<sup>3</sup> In this case, as one might expect, it is necessary to use pseudoinverses. As discussed in [11], there are various sets of properties one can impose in defining pseudoinverses. For our purposes here, it suffices to take the pseudoinverse  $Z^{\dagger}$  of a symmetric matrix Z to be any symmetric matrix for which

$$ZZ^{\dagger}Z = Z \tag{2.20}$$

(this is what is referred to in [11] as a (1)-inverse). Then we have the following.

Lemma 2.2: Suppose that H has full-column rank. The ML estimate of x based on the measurement vector (2.1)

$$\hat{x}_{\mathrm{ML}} = (0 \quad I) \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\mathsf{T}} \begin{pmatrix} I \\ 0 \end{pmatrix} y.$$
 (2.21)

This estimate is unbiased and has for error covariance

$$P_{\rm ML} = -\begin{pmatrix} 0 & I \end{pmatrix} \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\rm T} \begin{pmatrix} 0 \\ I \end{pmatrix}.$$
(2.22)

This result is proved in [11], although in our case we can say a bit more. Specifically, let

$$\begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\mathsf{T}} = \begin{pmatrix} W & U \\ U^T & T \end{pmatrix}.$$
 (2.23)

Then in the case where H has full-column rank, T is unique, while W and U will not be unless (2.10) is invertible. Note that from (2.21)

$$\hat{x}_{\mathsf{ML}} = U^T y \tag{2.24}$$

so that the gains in the ML estimator may not be unique --reflecting the fact that there are nonunique ways in which to determine certain linear combinations of components of x exactly. On the other hand, the resulting error covariance should be unambiguously defined, and from (2.22), (2.23) we see that it is, since  $P_{ML} = -T$ . We refer the reader to Appendix A for a summary of several of the calculations and some results from [11]. In particular, we prove the identity

$$\begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix} \begin{pmatrix} U \\ T \end{pmatrix} = \begin{pmatrix} 0 \\ I \end{pmatrix}$$
(2.25)

which is used below.

If H does not have full-column rank, then, as we have indicated, x is not estimable so that the ML estimate of all x is undefined. Nevertheless, in such a situation various linear combinations of x may be estimable [e.g., obviously  $(x_1 + x_2)$  is estimable from the observation y = $(x_1 + x_2)$ ]. The precise definition of estimability given in [11] is that the linear combination of  $c^T x$  is estimable if there exists a measurement linear combination  $d^T y$  that is an unbiased estimate of  $c^{T}x$ . An essentially immediate necessary and sufficient condition for this is that c must be in the range of  $H^T$ . Let r denote the rank of H and let  $H = H_1 H_2$  denote a full-rank factorization of H, i.e.,  $H_1$ is a  $p \times r$  full-column rank matrix and  $H_2$  is an  $r \times n$ full-row rank matrix. Then, it is precisely  $z = H_2 x$  whose ML estimate can be computed from y. Furthermore, from results in [11] we can deduce that

$$\hat{z}_{ML} = \begin{pmatrix} 0 & I \end{pmatrix} \begin{pmatrix} R & H_1 \\ H_1^T & 0 \end{pmatrix}^{\mathsf{T}} \begin{pmatrix} I \\ 0 \end{pmatrix} y$$
$$= \begin{pmatrix} 0 & H_2 \end{pmatrix} \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\mathsf{T}} \begin{pmatrix} I \\ 0 \end{pmatrix} y \qquad (2.26)$$

<sup>&</sup>lt;sup>3</sup> Indeed, while it may be easy to keep track of and eliminate redundancies in a given set of measurements, it is more difficult to do this in an organized and easily expressible way when those redundancies may evolve dynamically and arise through the dynamic constraints as well as the measurements

is an unbiased estimate of z with associated error covariance

$$P_{\rm ML} = -(0 \quad I) \begin{pmatrix} R & H_1 \\ H_1^T & 0 \end{pmatrix}^{\dagger} \begin{pmatrix} 0 \\ I \end{pmatrix}$$
$$= -(0 \quad H_2) \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\dagger} \begin{pmatrix} 0 \\ H_2^T \end{pmatrix} y \quad (2.27)$$

Next, we prove several results that provide the justification for the recursive procedure that will be employed below for computing ML estimates. The first of these results states that for the purpose of estimating other variables, we can replace several measurements of a variable by its estimate based on these measurements.

Lemma 2.3: Let x and z be unknown vectors and consider the observations

$$a = Hx + v \tag{2.28}$$

$$b = Jx + Kz + w \tag{2.29}$$

where v and w are independent, zero-mean Gaussian vectors with covariance matrices V and W, respectively. Suppose that x is estimable based on (2.28) only, and that z is estimable based on both (2.28) and (2.29). Let  $\hat{x}_1$  be the estimate of x based on (2.28), and let  $P_1$  be the associated error covariance matrix. Then  $\hat{x}$ , the estimate of x based on both (2.28) and (2.29), and its associated estimation error covariance P are identical to the estimate and estimation error covariance resulting from estimating x from (2.29) and the observation

$$\hat{x}_1 = x + u \tag{2.30}$$

where u is zero mean and Gaussian, independent of w, with covariance  $P_1$ .

Furthermore, the estimate  $\hat{z}$  and estimation error covariance of z based on (2.28) and (2.29) are the same as its estimate and error covariance based on (2.29) and (2.30).

**Proof:** Since x is estimable from (2.28), H must have full-column rank. This, coupled with the assumption that z is estimable from both (2.28) and (2.29), implies that K has full-column rank. Consider the joint estimation of x and z based on (2.28) and (2.29). If  $\lambda_a$  and  $\lambda_b$  denote the Lagrange multiplier vectors associated with observations (2.28) and (2.29), respectively, according to (2.9) the multiplier vectors and estimates  $\hat{x}$  and  $\hat{z}$  satisfy the system

$$\begin{pmatrix} V & 0 & H & 0 \\ 0 & W & J & K \\ H^T & J^T & 0 & 0 \\ 0 & K^T & 0 & 0 \end{pmatrix} \begin{pmatrix} \lambda_a \\ \lambda_b \\ \hat{x} \\ \hat{z} \end{pmatrix} = \begin{pmatrix} a \\ b \\ 0 \\ 0 \end{pmatrix}.$$
 (2.31)

We seek to compare this "batch" estimation method, where all the measurements are processed at the same time, with the "sequential" approach where we first estimate x based on (2.28) alone, and then combine the resulting estimate in the form of the summary measurement (2.30) with observation (2.29) in order to estimate

both x and z. If  $\lambda_{1a}$  and  $\hat{x}_1$  are the Lagrange multiplier vector and estimate of x based on (2.28) alone, and if  $U_1$  and  $P_1$  are the associated estimator matrix and error covariance, we see from (2.9) and from identity (2.25) that they satisfy

$$\begin{pmatrix} V & H \\ H^T & 0 \end{pmatrix} \begin{pmatrix} \lambda_{1a} & U_1 \\ \hat{x}_1 & -P_1 \end{pmatrix} = \begin{pmatrix} a & 0 \\ 0 & I \end{pmatrix}. \quad (2.32)$$

Similarly, the estimates of x and z based on (2.29) and the summary measurement (2.30) are obtained by solving

$$\begin{pmatrix} P_1 & 0 & I & 0 \\ 0 & W & J & K \\ I & J^T & 0 & 0 \\ 0 & K^T & 0 & 0 \end{pmatrix} \begin{pmatrix} \lambda_1 \\ \lambda_b \\ \hat{x} \\ \hat{z} \end{pmatrix} = \begin{pmatrix} \hat{x}_1 \\ b \\ 0 \\ 0 \end{pmatrix}$$
(2.33)

where  $\lambda_1$  denotes here the Lagrange multiplier vector associated with the measurement (2.30). What we need to show is that the solution of systems (2.32) and (2.33) combined is equivalent to solving (2.31). This requires eliminating  $\lambda_{1a}$ ,  $\hat{x}_1$ , and  $\lambda_1$  from (2.32) and (2.33) and finding an expression for the variable  $\lambda_a$  of (2.31) in terms of these quantities. Note that our choice of notation already indicates that the variables  $\hat{x}$ ,  $\hat{z}$ , and  $\lambda_b$  which appear in both systems are the same.

First, we observe that the second- and fourth-block row of (2.31) already appear as the second- and fourth-block rows of (2.33), so that we need only to obtain the first- and third-block rows of (2.31) from (2.32) and (2.33).

First-block row: We multiple the first-block row of (2.33) on the left by H. This gives

$$HP_1\lambda_1 + H\hat{x} = H\hat{x}_1. \tag{2.34}$$

Now using the identities corresponding to the (1,1) and (1,2) blocks of (2.32), we find

$$VU_1\lambda_1 + H\hat{x} = a - V\lambda_{1a}$$
(2.35)

so that if we denote

$$\lambda_a \triangleq \lambda_{1a} + U_1 \lambda_1 \tag{2.36}$$

we get the first-block row of (2.31).

Third-block row: Consider the third-block row of (2.33)

$$\lambda_1 + J^T \lambda_b = 0. \tag{2.37}$$

Substituting the identities corresponding to the (2,1) and (2,2) blocks of (2,32) inside this identity, we obtain

$$H_r^T \lambda_a + J_r^T \lambda_b = 0 \tag{2.38}$$

where  $\lambda_a$  is defined as in (2.36). But (2.38) is just the third-block row of (2.31).

Since the estimates of  $\hat{x}$  and  $\hat{z}$  obtained by the batch and sequential methods are identical functionals of a and b, and the statistical assumptions for a and b are the same under both methods, the error covariances are the same.

Note that in the above proof, the assumption that x is estimable from (2.28) was needed to ensure that the

second-block column of (2.32) holds. On the other hand, the assumption that z is estimable from both (2.28) and (2.29), i.e., that K has full rank, was only required insofar as we wanted to discuss the properties of the ML estimate  $\hat{z}$  and the associated error covariance. It is not necessary if we are only interested in estimable linear combinations of the entries of z, and the above result can easily be restated in a way that does not require z to be estimable.

Lemma 2.3 shows that previously processed measurements can be aggregated in the form of a summary measurement (2.30) for the variables that have been estimated. However, the summary measurement (2.30) will include estimates of variables that do not appear in subsequent measurements and in which we are no longer interested. Conventional wisdom suggests that measurements associated to such exogenous variables can be discarded without affecting the estimation of the other variables. The following result, which is expressed in its most general form, provides a criterion for dropping unneeded measurements.

Lemma 2.4: Consider the observations

$$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \begin{pmatrix} H_1 & 0 \\ H_2 & H_3 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} v_1 \\ v_2 \end{pmatrix}$$
(2.39)

where  $x_1$  and  $x_2$  are two unknown vectors, and  $[v_1^T \quad v_2^T]^T$  is a zero-mean Gaussian vector with covariance

$$\begin{pmatrix} R_{11} & R_{12} \\ R_{12}^T & R_{22} \end{pmatrix}.$$

Suppose that  $H_3$  has full-row rank, and  $H_1$  has fullcolumn rank. Then the ML estimate of  $x_1$  based on both  $y_1$  and  $y_2$  is the same as that based on  $y_1$  alone.

The assumption that  $H_1$  has full-column rank is introduced here to guarantee that  $x_1$  is estimable from the  $y_1$ measurement, but it can be removed if we only seek to estimate estimable linear combinations of the entries of  $x_1$ .

**Proof:** Let  $\lambda_1$  and  $\lambda_2$  be the Lagrange multiplier vectors associated with the  $y_1$  and  $y_2$  measurements, respectively, and let  $\hat{x}_1$  and  $\hat{x}_2$  be the estimates of  $x_1$  and  $x_2$  based on both  $y_1$  and  $y_2$ . According to (2.9), they satisfy the system

$$\begin{pmatrix} R_{11} & R_{12} & H_1 & 0\\ R_{12}^T & R_{22} & H_2 & H_3\\ H_1^T & H_2^T & 0 & 0\\ 0 & H_3^T & 0 & 0 \end{pmatrix} \begin{pmatrix} \lambda_1\\ \lambda_2\\ \hat{x}_1\\ \hat{x}_2 \end{pmatrix} = \begin{pmatrix} y_1\\ y_2\\ 0\\ 0 \end{pmatrix}. \quad (2.40)$$

Since  $H_3$  has full-row rank, the relation  $H_3^T \lambda_2 = 0$  implies that  $\lambda_2 = 0$ , so that we can delete the second- and fourth-block rows and columns from (2.40). This gives

$$\begin{pmatrix} R_{11} & H_1 \\ H_1^T & 0 \end{pmatrix} \begin{pmatrix} \lambda_1 \\ \hat{x}_1 \end{pmatrix} = \begin{pmatrix} y_1 \\ 0 \end{pmatrix}$$
(2.41)

which is precisely the system corresponding to the ML estimation of  $x_1$  (or estimable linear combinations of its entries) from  $y_1$  alone.

The combination of Lemmas 2.3 and 2.4 provides a general mechanism for generating ML estimates recursively. Specifically, consider the two observations

$$a = Hx_1 + Gx_2 + v$$
 (2.42)

$$b = Jx_1 + Kz + w \tag{2.43}$$

where we assume that  $x_1$  and  $x_2$  are jointly estimable from (2.42), i.e.,  $[H \ G]$  has full-column rank, and z is estimable from (2.42) and (2.43), i.e., K has full-column rank. As in Lemma 2.3, we assume that v and w are zero-mean independent Gaussian vectors with covariance V and W, respectively. From Lemma 2.3, we see that the measurement (2.42) can be replaced by the summary measurements

$$\hat{x}_1 = x_1 + u_1 \tag{2.44a}$$

$$\hat{x}_2 = x_2 + u_2 \tag{2.44b}$$

where  $\hat{x}_1$  and  $\hat{x}_2$  denote the ML estimates of  $x_1$  and  $x_2$ based on (2.42) alone, and the covariance of  $\begin{bmatrix} u_1^T & u_2^T \end{bmatrix}^T$  is the corresponding estimation error covariance. Then, the ML estimates of  $x_1$  and z based on both (2.43) and (2.44a) and (2.44b) are the same as those derived from (2.42) and (2.43). But  $x_2$  does not appear in observation (2.43), and the system obtained by combining (2.43) and (2.44a) and (2.44b) satisfies the assumptions of Lemma 2.4, so that for the purpose of estimating  $x_1$  and z we can drop the measurement (2.44b). This shows that the ML estimates of  $x_1$  and z based on both (2.43) and the summary measurement (2.44a) are the same as those based on (2.42) and (2.43). This general procedure, whereby previously estimated variables are replaced by summary measurements, and irrelevant variables are discarded, constitutes the basis for the descriptor Kalman filtering method of Section III.

Finally, we prove the intuitively obvious fact that for a given measurement set, if the noise covariance increases, the error covariance of the ML estimate also increases.

Lemma 2.5: Consider the observations

$$y_1 = Hx + v_1$$
 (2.45a)

$$y_2 = Hz + v_2 \tag{2.45b}$$

where  $v_1$  and  $v_2$  are zero-mean Gaussian vectors with covariances  $V_1$  and  $V_2$ , and suppose that H has full-column rank. Then, if  $V_2 \ge V_1$ , the estimation error variance associated with estimating x based on (2.45a) is less than or equal to the estimation error variance for estimating zbased on (2.45b).

*Proof:* Let  $\hat{x}_1$ ,  $P_1$  and  $\hat{z}_2$ ,  $P_2$  denote the ML estimates and estimation error covariances associated with (2.45a) and (2.45b), respectively.

As shown in Appendix A

$$\hat{z}_2 = U_2^T y_2, \qquad P_2 = U_2^T V_2 U_2$$
 (2.46)

where  $U_2^T$  is a left inverse of H satisfying some additional conditions. Consider then the following estimate of x:

$$\bar{x} = U_2^T y_1.$$
 (2.47)

Since  $U_2^T H = I$ , this is an unbiased estimate; however, it may be suboptimal. Thus

$$P_{1} \leq \mathscr{M} \Big[ (x - \bar{x})(x - \bar{x})^{T} \Big] = U_{2}^{T} V_{1} U_{2} \leq U_{2}^{T} V_{2} U_{2} = P_{2}.$$
(2.48)

### **III. THE DESCRIPTOR KALMAN FILTER**

We are now in a position to consider the recursive estimation problem for the system (1.1)-(1.3). As we have indicated, here we adopt an ML perspective, viewing the dynamics (1.1) and prior density on x(0) as additional measurements. Specifically, in this section we describe the recursive computation of the filtered estimate  $\hat{x}(j)$  which we define as the ML estimate of x(j) based on the "measurements" in (1.1) and (1.2) for  $k = 0, \dots, j - 1$ together with the "measurements" provided by the prior information about x(0)

$$\bar{x}_0 = x(0) + \nu \tag{3.1}$$

where  $\nu$  is zero-mean Gaussian and independent of u(k)and r(k), and with covariance  $P_0$ . Specifically, (3.1) and (1.1), (1.2) for  $k = 0, \dots, j - 1$  provide us with set measurements of the unkown vector  $[x^T(0), x^T(1), \dots, x^T(j)]$ . Examining this set of measurements we see that the only terms in these equations involving x(j) are of the form  $E_j x(j)$  and  $C_j x(j)$ . Thus, a necessary condition for x(j) to be estimable is that  $\begin{pmatrix} E_j \\ C_j \end{pmatrix}$  have full-column rank. By induction, using the recursive ML estimation procedure outlined in Section II, and the fact that x(0) is estimable from (3.1), we can show that this is also a sufficient condition for estimability and, in fact, we can establish the following.

Lemma 3.1: Let  $P_j$  denote the error covariance associated with the filtered estimate  $\hat{x}(j)$ , with  $\hat{x}(0) = \bar{x}_0$  and  $P_0$  given by the prior distribution for x(0). Then  $\bar{x}(j + 1)$  and  $P_{j+1}$  are, respectively, equal to the ML estimate of x(j + 1) and its associated estimation error covariance based on the following observations:

$$y(j+1) = C_{j+1}x(j+1) + r(j)$$
(3.2)

$$A_{j}\hat{x}(j) = E_{j+1}x(j+1) + A_{j}\nu(j) - u(j) \quad (3.3)$$

where  $\nu(j)$  is a Gaussian random vector, independent of r(j), with zero mean and variance  $P_j$ .

Applying Lemma 2.1 to (3.2), (3.3) provides us with the 3-block form of the descriptor Kalman filter summarized in the following.

Theorem 3.1: The filtered estimated  $\hat{x}(j+1)$  and the corresponding error variance  $P_{j+1}$  can be obtained from

the following recursions:

$$\hat{x}(j+1) = \begin{pmatrix} 0 & 0 & I \end{pmatrix} \begin{pmatrix} A_j P_j A_j^T + Q_j & -S_j & E_{j+1} \\ -S_j^T & R_j & C_{j+1} \\ E_{j+1}^T & C_{j+1}^T & 0 \end{pmatrix}^{\dagger} \\ \cdot \begin{pmatrix} A_j \hat{x}(j) \\ y(j+1) \\ 0 \end{pmatrix} \qquad (3.4) \\ \begin{pmatrix} A_j P_j A_j^T + Q_j & -S_j & E_{j+1} \end{pmatrix}^{\dagger}$$

$$P_{j+1} = \begin{pmatrix} 0 & 0 & I \end{pmatrix} \begin{pmatrix} n_j n_j n_j + 2j & 5j & 2j+1 \\ -S_j^T & R_j & C_{j+1} \\ E_{j+1}^T & C_{j+1}^T & 0 \end{pmatrix} \cdot \begin{pmatrix} 0 \\ 0 \\ I \end{pmatrix}.$$
(3.5)

If past and present observations and dynamics do not supply redundant perfect information, i.e., when

$$\begin{pmatrix} A_j P_j A_j^T + Q_j & S_j & E_{j+1} \\ S_i^T & R_j & C_{j+1} \end{pmatrix}$$

has full-row rank, then the pseudoinverse in (3.4) and (3.5) is, in fact, an inverse.

Equation (3.4) is the 3-block form of the optimal filter, with the Hamiltonian matrix being the 3-block matrix whose pseudoinverse is taken on the right-hand side. Van Dooren [32] introduced a similar Hamiltonian form in the case in which E = I and R is invertible (so the Hamiltonian matrix is certainly invertible) and, by a limiting argument suggested that it was also valid for R singular. Whittle [34, chapter 11] essentially shows this using a dual optimization approach as in (2.7)-(2.9) but applied to the dynamic problem directly. Arnold and Laub [3] also consider such a form for R invertible and  $E \neq I$  but invertible. Perhaps closer to our work is that of Mehrmann [22] who considers a related Hamiltonian pencil (which we encounter in Section V) for the case of E and R singular in the context of an optimal control problem for singular systems. However, no recursion of the type (3.4) is presented in [22] and, even more importantly, the 3-block descriptor Riccati equation (3.5) is not obtained in any of these references. For completeness, we note that Bender and Laub [4], [5] do introduce a Riccati equation for the case of E singular in the context of an optimal control problem. However, theirs is a reduced-order Riccati equation of size equal to the rank of E, rather than the full-order Riccati equation  $(3.5)^4$ .

Finally, we remark that the previously derived results referred to in the introduction are indeed specializations of (3.4) and (3.5). Wang and Bernhard [33] examine the case of time-invariant systems, where it is assumed not

<sup>&</sup>lt;sup>4</sup> For this reason we conjecture that there is a connection between the estimation dual of Bender and Laub [4], [5] and the work in [33] in which the focus is on estimating Ex(k). This is only of tangential interest here and thus is not pursued.

only that  $\begin{pmatrix} E \\ C \end{pmatrix}$  has full-column rank, but also that  $\begin{pmatrix} E \\ C \end{pmatrix}$  has full-row rank, S = 0, and R is invertible. Their analysis can be interpreted as estimating Ex(j) = Ax(j-1) + u(j-1) based on observations y(k) for  $0 \le k \le j-1$ . In this case, they obtain the following Riccati equation for the error covariance  $\Sigma_j$  of this one-step predicted estimate:

$$\Sigma_{j+1} = (0 \quad A) M_j^{-1} \begin{pmatrix} \Sigma_j & 0 \\ 0 & C^T R^{-1} C \end{pmatrix} M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} + Q$$
(3.6)

with

$$M_j = \begin{pmatrix} \Sigma_j & E \\ E^T & -C^T R^{-1} C \end{pmatrix}.$$
 (3.7)

To see how (3.6) and (3.7) can be obtained from (3.5), note that

$$\Sigma_i = AP_i A^T + Q \tag{3.8}$$

and thus using (3.5) and the fact that in Wang and Bernhard's case past and present observations and dynamics do not supply redundant perfect information, we obtain the following recursion for  $\Sigma_i$ :

$$\Sigma_{j+1} = -(0 \quad 0 \quad A) \begin{pmatrix} R & 0 & C \\ 0 & \Sigma_j & E \\ C^T & E^T & 0 \end{pmatrix}^{-1} \begin{pmatrix} 0 \\ 0 \\ A^T \end{pmatrix} + Q.$$
(3.9)

Then the use of the standard block-matrix inversion results allows us to express (3.9) as

$$\Sigma_{j+1} = -(0 \quad A)M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} + Q$$
 (3.10)

where  $M_j$  is defined as in (3.7). It turns out that (3.10) is a simplified version of (3.6). To obtain (3.10) from (3.6), note that

$$\begin{split} \Sigma_{j+1} &= (0 \quad A) M_j^{-1} \begin{pmatrix} \Sigma_j & 0 \\ 0 & C^T R^{-1} C \end{pmatrix} M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} + Q \\ &= (0 \quad A) M_j^{-1} \begin{bmatrix} M_j - \begin{pmatrix} 0 & E \\ E^T & 2C^T R^{-1} C \end{pmatrix} \end{bmatrix} M_j^{-1} \\ &\cdot \begin{pmatrix} 0 \\ A^T \end{pmatrix} + Q \\ &= (0 \quad A) M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} - (0 \quad A) M_j^{-1} \\ &\cdot \begin{pmatrix} 0 & E \\ 0 & C^T R^{-1} C \end{pmatrix} M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} \\ &- (0 \quad A) M_j^{-1} \begin{pmatrix} 0 \\ E^T & C^T R^{-1} C \end{pmatrix} M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} \\ &= (0 \quad A) M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} - (0 \quad A) \begin{pmatrix} 0 \\ 0 \end{bmatrix} M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} + Q \\ &= (0 \quad A) M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} - (0 \quad A) \begin{pmatrix} 0 \\ 0 \end{bmatrix} M_j^{-1} \begin{pmatrix} 0 \\ A^T \end{pmatrix} \end{split}$$

$$-(0 \quad A)M_{j}^{-1}\begin{pmatrix} 0 & 0 \\ 0 & I \end{pmatrix} \begin{pmatrix} 0 \\ A^{T} \end{pmatrix} + Q$$
  
=  $-(0 \quad A)M_{j}^{-1}\begin{pmatrix} 0 \\ A^{T} \end{pmatrix} + Q.$  (3.11)

If in addition,  $\Sigma_i$  is invertible, we can express the inverse of  $M_i$  in (3.10) in terms of its block entries, so that

$$\Sigma_{j+1} = A \Big( E^T \Sigma_j^{-1} E + C^T R^{-1} C \Big)^{-1} A^T + Q \quad (3.12)$$

which is the Riccati equation that we had considered earlier in [28]. In the nondescriptor case (E = I), (3.12) reduces to the standard Riccati equation of Kalman filtering

$$\Sigma_{j+1} = A \left( \Sigma_j^{-1} + C^T R^{-1} C \right)^{-1} A^T + Q. \quad (3.13)$$

IV. STABILITY AND CONVERGENCE OF THE DESCRIPTOR KALMAN FILTER

In this section, we study the asymptotic properties of the descriptor Kalman filter in the time-invariant case

$$Ex(k+1) = Ax(k) + u(k), \quad k \ge 0 \quad (4.1)$$
  
$$y(k+1) = Cx(k+1) + r(k), \quad k \ge 0 \quad (4.2)$$

where matrices E and A are  $l \times n$ , C is  $p \times n$ , and u and r are zero mean, white, Gaussian sequences with covariance

$$\mathscr{M}\left[\binom{u(k)}{r(k)}\binom{u(j)}{r(j)}^{T}\right] = \binom{Q}{S^{T}} \binom{S}{R}\delta(k-j). \quad (4.3)$$

In particular, the results presented in this section generalize those in [23], [28] and [33] for descriptor systems and the usual results for standard causal systems. Note that as in [33] in our development we do *not* require that l = n, so that (4.1) need not be square and even if it is, we do not require  $\{E, A\}$  to define a regular pencil. In the context of viewing (4.1) as simply providing another source of measurements, we see that it is quite natural to remove this restriction. Also, as before, we allow the possibility that R is singular as well.

Definition 4.1: The system (4.1), (4.2) is called detectable if

$$\begin{pmatrix} sE - tA \\ C \end{pmatrix}$$

has full-column rank for all  $(s, t) \neq (0, 0)$  such that  $|s| \ge |t|$ . It is called stabilizable if

$$\begin{pmatrix} sE - tA & Q & -S \\ sC & -S^T & R \end{pmatrix}$$

has full-row rank for all  $(s, t) \neq (0, 0)$  such that  $|s| \ge |t|$ .

These definitions generalize other similar definitions in the literature [7], [22]. Note, for example, that these definitions reduce to the classical notions of detectability and stabilizability when E = I and R > 0.5

<sup>5</sup> In this case our detectability condition reduces to  $\binom{sI-A}{C}$  having full-column rank for all |s| > 1. Also, with R > 0 and since (4.3) implies that  $Q = BB^T$  and  $S = BD^T$  for some matrices B and D, stabilizability in this case corresponds to  $(sI - A \ B)$  having full-row rank for |s| > 1.

The following generalizes the usual result that detectability implies the existence of a stable observer. An important point to note here, however, is that our observer (4.4) is an explicit causal system, even if (4.1), (4.2)are implicit, a direct consequence of estimability.

Theorem 4.1: Let (4.1), (4.2) be detectable, then there exists a stable filter

$$x_{s}(k+1) = A_{s}x_{s}(k) + K_{s}y(k+1), \qquad x_{s}(0) = \bar{x}_{0}$$
(4.4)

i.e., such that  $A_s$  is a stable matrix (all eigenvalues have magnitude less than one) and with

$$\lim_{k\to\infty} \mathscr{M}\Big[(x(k)-x_s(k))(x(k)-x_s(k))^T\Big] < \infty. \quad (4.5)$$

*Proof of Theorem 4.1:* We start the proof by showing the following lemma.

Lemma 4.1: Let (4.1), (4.2) be detectable. Then there exists a left inverse  $(L_e \ L_c)$  of  $\begin{pmatrix} E \\ C \end{pmatrix}$ , i.e.,

$$L_e E + L_c C = I \tag{4.6}$$

such that  $L_{e}A$  is stable.

**Proof of Lemma 4.1:** First note that the lemma is trivially true if C has full-column rank, since we can simply take  $L_e = 0$  and  $L_c = a$  left inverse of C. Assuming this is not the case but that  $\begin{pmatrix} E \\ C \end{pmatrix}$  has full rank, we can find invertible  $l \times l$  and  $n \times n$  matrices U and V such that

$$UEV = \begin{pmatrix} I & 0\\ 0 & E_{22} \end{pmatrix}$$
(4.7)

$$CV = \begin{pmatrix} 0 & C_2 \end{pmatrix} \tag{4.8}$$

where the partitions in (4.7), (4.8) are compatible, I denotes a square identity matrix, and  $C_2$  has full-column rank. If we partition UAV similarly as

$$UAV = \begin{pmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{pmatrix}$$
(4.9)

the detectability of (C, E, A) implies that

$$\begin{pmatrix} (s/t)I - A_{11} \\ -A_{21} \end{pmatrix}$$

has full-column rank for  $|s/t| \ge 1$ , which means that  $(A_{11}, -A_{21})$  is detectable in the usual sense. Thus, there exists a matrix D such that  $A_{11} + DA_{21}$  is stable. Next, let F be any matrix satisfying

$$FC_2 = \begin{pmatrix} -DE_{22} \\ I \end{pmatrix}.$$
 (4.10)

For example, we can take

$$F = \begin{pmatrix} -DE_{22}C_2^L \\ C_2^L \end{pmatrix}.$$
 (4.11)

where 
$$C_2^L$$
 is any left inverse of  $C_2$ . Then

$$\begin{pmatrix} I & D \\ 0 & 0 \end{pmatrix} \begin{pmatrix} I & 0 \\ 0 & E_{22} \end{pmatrix} + F(0 \quad C_2) = I \quad (4.12)$$

and

$$\begin{pmatrix} I & D \\ 0 & 0 \end{pmatrix} \begin{pmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{pmatrix} = \begin{pmatrix} A_{11} + DA_{21} & A_{12} + DA_{22} \\ 0 & 0 \end{pmatrix}$$
(4.13)

which is stable because  $A_{11} + DA_{22}$  is stable. Thus, by taking

$$L_e = V \begin{pmatrix} I & D \\ 0 & 0 \end{pmatrix} U \tag{4.14a}$$

$$L_c = VF \tag{4.14b}$$

the lemma is proved.

Continuing the proof of the theorem, note that using the above lemma, we can express x(k + 1) as

$$x(k+1) = L_e A x(k) + L_c y(k+1) + L_e u(k) - L_c r(k)$$
(4.15)

where  $L_{e}A$  is stable. If we now define

$$x_s(k+1) = L_e A x_s(k) + L_c y(k+1)$$
 (4.16)

we can easily see that

$$\lim_{k\to\infty} \mathscr{M}\left[(x(k)-x_s(k))(x(k)-x_s(k))^T\right] = P_s \quad (4.17)$$

where  $P_s$  is the unique positive semidefinite solution of the Lyapunov equation

$$P_{s} - (L_{e}A)P_{s}(L_{e}A)^{T} = \begin{pmatrix} L_{e} & L_{c} \end{pmatrix} \begin{pmatrix} Q & -S \\ -S^{T} & R \end{pmatrix} \begin{pmatrix} L_{e}^{T} \\ L_{c}^{T} \end{pmatrix}.$$
(4.18)

The theorem is thus proved.

Detectability alone, of course, does not guarantee that the descriptor Kalman filter converges to a stable filter. However, as would also be expected from what we know for causal systems, detactability does tell us something about the descriptor algebraic Riccati equation.

Theorem 4.2: Let (4.1), (4.2) be detectable. Then the algebraic descriptor Riccati equation

$$P = -(0 \ 0 \ I) \begin{pmatrix} APA^{T} + Q \ -S \ E \\ -S^{T} \ R \ C \\ E^{T} \ C^{T} \ 0 \end{pmatrix}^{\dagger} \begin{pmatrix} 0 \\ 0 \\ I \end{pmatrix} (4.19)$$

has a positive semidefinite solution.

*Proof:* We prove the existence of a positive semidefinite solution P to (4.19) by showing that the descriptor

• \_,

Riccati recursion

$$P_{j+1} = -(0 \quad 0 \quad I) \begin{pmatrix} AP_{j}A^{T} + Q & -S & E \\ -S^{T} & R & C \\ E^{T} & C^{T} & 0 \end{pmatrix}^{\dagger} \begin{pmatrix} 0 \\ 0 \\ I \end{pmatrix}$$
(4.20)

with  $P_0 = 0$  is monotone increasing and bounded. This of course implies the convergence of  $P_k$ , which, from (4.20) implies that this limit, which must be positive semidefinite, satisfies (4.19).6

To see the boundedness of  $P_k$ , consider the stable filter (4.16) with  $x_s(0) = \bar{x}_0$ . It is then clear that the associated error variance matrices  $P_s(k)$  converge asymptotically to  $P_s$  the unique solution of (4.18), and that thanks to the optimality of the descriptor Kalman filter,  $P_k \leq P_s(k)$ .

We show that  $P_k$  is monotone increasing by induction. Clearly

$$P_1 \ge P_0 = 0. \tag{4.21}$$

Now suppose that

$$P_j \ge P_{j-1}.\tag{4.22}$$

 $P_i$  is the estimation error covariance associated with estimating x(j) based on

$$\begin{pmatrix} A\hat{x}(j-1) \\ y(j) \end{pmatrix} = \begin{pmatrix} E \\ C \end{pmatrix} x(j) + \begin{pmatrix} A\nu(j-1) - u(j-1) \\ r(j-1) \end{pmatrix}$$
(4.23)

where the covariance of  $\begin{pmatrix} A\nu(j-1) - u(j-1) \\ r(j-1) \end{pmatrix}$  is  $\begin{pmatrix} AP_{j-1}A^T + Q & -S \\ -S^T & R \end{pmatrix}$ . Also  $P_{j-1}$  is the estimation error

covariance associated with estimating x(i + 1) based on

$$\begin{pmatrix} A\hat{\mathbf{x}}(j) \\ \mathbf{y}(j+1) \end{pmatrix} = \begin{pmatrix} E \\ C \end{pmatrix} \mathbf{x}(j+1) + \begin{pmatrix} A\nu(j) - u(j) \\ r(j) \end{pmatrix}$$
(4.24)

where the covariance of  $\begin{pmatrix} A\nu(j) - u(j) \\ r(j) \end{pmatrix}$  is  $\begin{pmatrix} AP_jA^T + Q & -S \\ -S^T & R \end{pmatrix}$ . But from (4.22) we see that  $\begin{pmatrix} AP_{j}A^{T} + Q & -S \\ -S^{T} & R \end{pmatrix} \ge \begin{pmatrix} AP_{j-1}A^{T} + Q & -S \\ -S^{T} & R \end{pmatrix}$ (4.25)

and thanks to Lemma 2.5, we conclude that  $P_{i+1} \ge P_i$ .

<sup>6</sup> The usual argument here involves a right-hand side which includes matrix inverses for which we can deduce convergence by the continuity of the inversion operation. While the full generalized inverse in (4.20) is not unique, the lower right-hand block is, due to the fact that  $\begin{pmatrix} E \\ C \end{pmatrix}$  has full-column rank. Indeed in Appendix A we give an explicit form for this block which involves true inverses (identify  $\begin{pmatrix} E \\ C \end{pmatrix}$  with H in Appendix A).

In the following, it is useful to have available an alternate form for the Kalman filter equation (3.4), (3.5) in the time-invariant case. Obtaining this requires the following which is proved in Appendix A:

Lemma 4.2: Let R be a positive semidefinite matrix and H a full-column rank matrix. Then

$$(0 \quad I)\left[\begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\dagger} + \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\dagger}\begin{pmatrix} R & 0 \\ 0 & 0 \end{pmatrix} \cdot \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\dagger}\right]\begin{pmatrix} 0 \\ I \end{pmatrix} = 0 \quad (4.26)$$

and

$$\begin{pmatrix} 0 & I \end{pmatrix} \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}^{\dagger} \begin{pmatrix} H \\ 0 \end{pmatrix} = I.$$
(4.27)

Using this result, it is straightforward to verify that the Riccati recursion (4.20) can be rewritten as

$$P_{j+1} = (L_j A) P_j (L_j A)^T + \begin{pmatrix} L_j & K_j \end{pmatrix} \begin{pmatrix} Q & -S \\ -S^T & R \end{pmatrix} \begin{pmatrix} L_j^T \\ K_j^T \end{pmatrix}$$
(4.28)

with

$$\begin{pmatrix} L_j & K_j \end{pmatrix} = \begin{pmatrix} 0 & 0 & I \end{pmatrix} \begin{pmatrix} AP_j A^T + Q & -S & E \\ -S^T & R & C \\ E^T & C^T & 0 \end{pmatrix}^{\dagger} \cdot \begin{pmatrix} I & 0 \\ 0 & I \\ 0 & 0 \end{pmatrix}.$$
(4.29)

Thus, we have that the descriptor Kalman filter (3.4) takes the form

$$\hat{x}(k+1) = L_k A \hat{x}(k) + K_k y(k+1).$$
 (4.30)

Via similar manipulations we can also rewrite the algebraic Riccati equation (4.19) as

$$P = (LA)P(LA)^{T} + (L \quad K)\begin{pmatrix} Q & -S \\ -S^{T} & R \end{pmatrix} \begin{pmatrix} L^{T} \\ K^{T} \end{pmatrix}$$
(4.31)

$$(L \ K) = (0 \ 0 \ I) \begin{pmatrix} APA^{T} + Q & -S & E \\ -S^{T} & R & C \\ E^{T} & C^{T} & 0 \end{pmatrix}^{'} \begin{pmatrix} I & 0 \\ 0 & I \\ 0 & 0 \end{pmatrix}.$$
(4.32)

In the following, we consider the behavior of the Kalman filter and Riccati equation when the system is both detectable and stabilizable, obtaining a generalization of well-known results for causal systems. Note that stabilizability implies that

$$\begin{pmatrix} E & Q & S \\ C & S^T & R \end{pmatrix}$$

has full-row rank, which in turn implies that

$$\begin{pmatrix} E & AP_jA^T + Q & S \\ C & S^T & R \end{pmatrix}$$

has full-row rank for all  $P_j \ge 0$ , so that in this case the pseudoinverses in (4.29), (4.32) are in fact inverses.

Theorem 4.3: Suppose that (4.1), (4.2) is detectable and stabilizable. Then for any initial condition  $P_0$ , the solution  $P_k$  of the Riccati equation

$$P_{k+1} = (L_k A) P_k (L_k A)^T + (L_k K_k) \begin{pmatrix} Q & -S \\ -S^T & R \end{pmatrix} \begin{pmatrix} L_k^T \\ K_k^T \end{pmatrix} (4.33)$$

$$(L_k K_k) = (0 \quad 0 \quad I) \begin{pmatrix} AP_k A^T + Q & -S & E \\ -S^T & R & C \\ E^T & C^T & 0 \end{pmatrix}^{\dagger} \cdot \begin{pmatrix} I & 0 \\ 0 & I \\ 0 & 0 \end{pmatrix}. (4.34)$$

converges exponentially fast to the unique positive semidefinite solution of the algebraic descriptor Riccati equation

$$P = (LA)P(LA)^{T} + (L \quad K)\begin{pmatrix} Q & -S \\ -S^{T} & R \end{pmatrix}\begin{pmatrix} L^{T} \\ K^{T} \end{pmatrix}$$
(4.35)

$$(L \quad K) = (0 \quad 0 \quad I) \begin{pmatrix} APA^{T} + Q & -S & E \\ -S^{T} & R & C \\ E^{T} & C^{T} & 0 \end{pmatrix}^{\dagger} \cdot \begin{pmatrix} I & 0 \\ 0 & I \\ 0 & 0 \end{pmatrix}. \quad (4.36)$$

Furthermore, the steady-state Kalman filter

$$\hat{x}(k+1) = LA\hat{x}(k) + Ky(k+1).$$
 (4.37)

is stable.

**Proof:** From Theorem 4.2, we know that there is at least one positive semidefinite solution to (4.35), (4.36). What we would like to show is that this positive semidefinite solution is unique, that  $P_j$  in (4.33), (4.34) converges to P exponentially fast for any initial condition  $P_0$ , and that the resulting steady-state Kalman filter (4.37) is exponentially stable, i.e., that LA is a stable matrix. Note that by using (4.27), it is not difficult to show that

$$LE = I - KC. \tag{4.38}$$

Thus, premultiplying (4.1) by L, using (4.2) and (4.37), and defining  $\tilde{x}(k) = x(k) - \hat{x}(k)$ , we see that

$$\tilde{x}(k+1) = LA\tilde{x}(k) + Lu(k) - Kr(k)$$
 (4.39)

so that this will imply the stability of the error dynamics.

Let us first show that LA is stable when P is taken, as any positive semidefinite solution of (4.31). Suppose LA is not stable. Then there exist a complex number  $\lambda$  and a complex row vector v such that  $|\lambda| \ge 1$  and

$$\nu L A = \lambda \nu. \tag{4.40}$$

From (4.31) we got that

$$(1 - |\lambda|^2) v P v^H = v (L - K) \begin{pmatrix} Q & S \\ S^T & R \end{pmatrix} \begin{pmatrix} L^T \\ K^T \end{pmatrix} v^H \quad (4.41)$$

where  $(\cdot)^{H}$  denotes the conjugate-transpose. Since the right-hand side of (4.41) is nonnegative and its left-hand side, nonpositive, we must have

$$\upsilon(L-K)\begin{pmatrix} Q-S\\S^T-R \end{pmatrix}=0.$$
(4.42)

But from (4.38) and (4.40) we have

$$\lambda v L E = v L A - \lambda v K C. \qquad (4.43)$$

From (4.42) and (4.43) it follows that

$$v(L - K) \begin{pmatrix} \lambda E - A & Q & S \\ \lambda C & S^T & R \end{pmatrix} = 0.$$
(4.44)

which since  $v(L - K) \neq 0$  [(4.40) implies  $vL \neq 0$ ] contradicts the stabilizability assumption. Thus, LA is stable.

Next we can show that there exists a unique positive semidefinite solution of (4.31). Specifically, suppose that  $P^1$  and  $P^2$  are two such solutions, and let  $[L^1 \ K^1], [L^2 \ K^2]$  denote the corresponding matrices in (4.32). Then  $L^1 A$  and  $L^2 A$  both are stable, and, as shown in Appendix B

$$P^{1} - P^{2} = (L^{2} A)(P^{1} - P^{2})(L^{1} A)^{T}$$
 (4.45)

so that iterating (4.45)

$$P^{1} - P^{2} = \lim_{k \to \infty} \left( L^{2} A \right)^{k} \left( P^{1} - P^{2} \right) \left[ \left( L^{1} A \right)^{T} \right]^{k} = 0.$$
(4.46)

Finally, we can show that  $P_j$  converges to P exponentially fast for any initial condition  $P_0$ . First note that  $P_j^0 \leq P_j$  where  $P_j^0$  is the error covariance for the problem starting from  $P_0 = 0$ . We already know that  $P_j^0 \rightarrow P$ . Thus, if we can find a sequence  $W_j$  so that  $P_j \leq W_j$  and  $W_j \rightarrow P$  exponentially fast, we will be finished. We accomplish this by letting  $W_j$  be the error covariance of the estimator defined by the steady-state filter (4.37) for all k, starting with the same initial estimate as the optimal Kalman filter. Thus,  $W_0 = P_0$  and  $W_j \geq P_j$  for all  $j \geq 0$ . Furthermore, from (4.39)

$$W_{j+1} = (LA)W_j(LA)^T + (L-K)\begin{pmatrix} Q & -S \\ -S^T & R \end{pmatrix} \begin{pmatrix} L \\ K \end{pmatrix}$$
(4.47)

which, thanks to the stability of LA, converges exponentially fast to the unique positive semidefinite solution of

$$W = (LA)W(LA)^{T} + (L \quad K) \begin{pmatrix} Q & -S \\ -S^{T} & R \end{pmatrix} \begin{pmatrix} L \\ K \end{pmatrix}$$
(4.48)

and a comparison to (4.31) yields that W = P, completing the proof.

Again we note that the results of this section represent a generalization of those in [28], [33]. Furthermore, they represent what to our knowledge is a new derivation for the more frequently studied singular estimation problem (E = I, R singular).

## V. CONSTRUCTION OF THE STEADY-STATE FILTER

In this section, we show that the solution of the algebraic descriptor Riccati equation can be constructed by using the eigenvectors and generalized eigenvectors of the pencil

$$\left\{ \begin{pmatrix} E & -Q & S \\ C & S^T & -R \\ 0 & A^T & 0 \end{pmatrix}, \begin{pmatrix} A & 0 & 0 \\ 0 & 0 & 0 \\ 0 & E^T & C^T \end{pmatrix} \right\}.$$
 (5.1)

A similar pencil was also introduced in [22] for the study of the LQ control problem for descriptor systems, although 3-block Riccati equations are neither introduced nor studied.

Throughout this section we shall assume that the system is detectable and stabilizable. The results we present here generalize the usual results [17], [31], [32] for standard causal (E = I) systems for the case of singular measurement noise, and our results of [23], [28] to the case where R may be singular and in addition  $\{E, A\}$ need not be regular nor even square. Before beginning, let us introduce the following notation:

$$F = \begin{pmatrix} E \\ C \end{pmatrix}, K = \begin{pmatrix} A \\ 0 \end{pmatrix}, G = \begin{pmatrix} Q & -S \\ -S^T & R \end{pmatrix}.$$
 (5.2)

The pencil (5.1) can now be expressed as

$$\left\{ \begin{pmatrix} F & -G \\ 0 & K^T \end{pmatrix}, \begin{pmatrix} K & 0 \\ 0 & F^T \end{pmatrix} \right\}.$$
 (5.3)

and the descriptor Riccati equation as

$$P = -\begin{pmatrix} 0 & I \end{pmatrix} \begin{pmatrix} KPK^T + G & F \\ F^T & 0 \end{pmatrix}^{-1} \begin{pmatrix} 0 \\ I \end{pmatrix}. \quad (5.4)$$

We begin with the following.

Lemma 5.1: The pencil (5.1) is regular and has no eigenmode on the unit circle.

*Proof:* All we need to show is that for all z on the unit circle

$$\begin{pmatrix} F & -G \\ 0 & K^T \end{pmatrix} + z \begin{pmatrix} K & 0 \\ 0 & F^T \end{pmatrix}$$
(5.5)

is invertible. Note that thanks to the detectability assumption which can now be stated in terms of the new notation as: "sF - tK has full-column rank for  $(s, t) \neq (0, 0)$  and  $|s| \ge |t|$ ," we can see that F + zK has full-column rank for all z on the unit circle. Now suppose that (5.5) is not invertible, which means that there exist u and v not simultaneously null such that

$$\begin{pmatrix} F+zK & -G\\ 0 & K^T+zF^T \end{pmatrix} \begin{pmatrix} u\\ v \end{pmatrix} = 0.$$
 (5.6)

If we now let

$$\Gamma = zK + F \tag{5.7}$$

from (5.6) it follows that

$$\Gamma u - Gv = 0 \tag{5.8}$$

$$\Gamma^{H} v = 0. \tag{5.9}$$

If we now multiply (5.8) and (5.9) on the left by  $v^{H}$  and  $u^{H}$ , respectively, and take the transpose conjugate of (5.9) and subtract from (5.8), we get

$$vGv^H = 0 \tag{5.10}$$

which since G is symmetric positive semidefinite implies that Gv = 0. Thus, since  $\Gamma$  has full-column rank, (5.8) implies that u = 0. But we also have that  $v^H(\Gamma \cap G) = 0$ which thanks to the stabilizability assumption implies v =0, contradicting the assumption that u and v are not simultaneously null.

Lemma 5.2: The pencil (5.1) has exactly *n* stable eigenmodes.

Proof: Let

$$p(s,t) = \det\left(s\left(\begin{matrix} F & -G\\ 0 & K^{T} \end{matrix}\right) + t\left(\begin{matrix} K & 0\\ 0 & F^{T} \end{matrix}\right)\right)$$
$$= \det\left(s\left(\begin{matrix} 0 & K^{T}\\ F & -G \end{matrix}\right)$$
$$+ t\left(\begin{matrix} 0 & F^{T}\\ K & 0 \end{matrix}\right)\right)$$
$$= \det\left(\begin{matrix} 0 & sK^{T} + tF^{T}\\ sF + tK & -sG \end{matrix}\right).$$
(5.11)

Then

$$p(t,s) = \det\left(t\begin{pmatrix}0 & K^{T}\\F & -G\end{pmatrix} + s\begin{pmatrix}0 & F^{T}\\K & 0\end{pmatrix}\right)$$
$$= \det\left(\begin{array}{cc}0 & tK^{T} + sF^{T}\\tF + sK & -tG\end{array}\right).$$
(5.12)

From

$$\begin{pmatrix} I & 0 \\ 0 & tI \end{pmatrix} \begin{pmatrix} 0 & sK^T + tF^T \\ sF + tK & -sG \end{pmatrix} \begin{pmatrix} I/t & 0 \\ 0 & I \end{pmatrix}$$

$$= \begin{pmatrix} I & 0 \\ 0 & sI \end{pmatrix} \begin{pmatrix} 0 & tK^T + sF^T \\ tF + sK & -tG \end{pmatrix}^T \begin{pmatrix} I/s & 0 \\ 0 & I \end{pmatrix}$$
(5.13) we find

we find

$$t^{l+p}p(s,t)t^{-n} = s^{l+p}p(t,s)s^{-n}$$
 (5.14)

so that

$$t^{l+p-n}p(s,t) = s^{l+p-n}p(t,s).$$
(5.15)

If we denote the number of zero eigenmodes by  $\delta_0$ , stable but nonzero eigenmodes by  $\delta_s$ , unstable eigenmodes by  $\delta_u$ , and infinite eigenmodes by  $\delta_\infty$ , from (5.15) and the fact that there are no eigenmodes on the unit circle, we conclude that

$$\delta_s = \delta_u \tag{5.16a}$$

$$\delta_{\infty} - \delta_0 = l + p - n. \tag{5.16b}$$

Finally noting that

$$\delta_0 + \delta_s + \delta_u + \delta_\infty = n + l + p \qquad (5.17)$$

we get that the number of stable eigenmodes  $\delta_0 + \delta_s = n$ .

Theorem 5.1: Let the columns of

 $\begin{pmatrix} X \\ Y_1 \\ Y_2 \end{pmatrix}$ 

form a basis for the eigenspace of the pencil (5.1) associated with its n stable eigenmodes, i.e.,

$$\begin{pmatrix} E & -Q & -S \\ C & -S^{T} & -R \\ 0 & A^{T} & 0 \end{pmatrix} \begin{pmatrix} X \\ Y_{1} \\ Y_{2} \end{pmatrix} = \begin{pmatrix} A & 0 & 0 \\ 0 & 0 & 0 \\ 0 & E^{T} & C^{T} \end{pmatrix} \begin{pmatrix} X \\ Y_{1} \\ Y_{2} \end{pmatrix}$$
(5.18)

where J is stable. Then, P, the unique positive semidefinite solution of the algebraic Riccati equation (4.19) is given by

$$P = X(E^T Y_1 + C^T Y_1)^{-1}.$$
 (5.19)

Proof: Using notation (5.2) and letting

$$Y = \begin{pmatrix} Y_1 \\ Y_2 \end{pmatrix}$$
(5.20)

we must show that

$$P = X(FY)^{-1}.$$
 (5.21)

To construct a real basis

$$\begin{pmatrix} X \\ Y \end{pmatrix}$$

and a real stable matrix J satisfying (5.18), we need only to compute the generalized real Schur decomposition [15, p. 396]

$$Q^{T} \begin{pmatrix} F & -G \\ 0 & K^{T} \end{pmatrix} Z = M$$
 (5.22a)

$$Q^{T} \begin{pmatrix} K & 0\\ 0 & F^{T} \end{pmatrix} Z = N$$
 (5.22b)

of the pencil (5.1), where Q and Z are orthogonal matrices, M is quasi-upper triangular, N is upper triangular, and where the  $n \times n$  blocks  $M_s$  and  $N_s$  in the decomposi-

tion

so that

$$M = \begin{pmatrix} M_s & M_{s\hat{s}} \\ 0 & M_{\hat{s}} \end{pmatrix}, N = \begin{pmatrix} N_s & N_{s\hat{s}} \\ 0 & N_2 \end{pmatrix}$$
(5.23)

correspond to the stable eigenmodes of the pencil (5.1), i.e.,  $J = N_s^{-1}M_s$  is stable. Then, if  $Z_s$  is the matrix formed by the first *n* columns of Z we have

$$Z_s = \begin{pmatrix} X \\ Y \end{pmatrix}. \tag{5.24}$$

From (5.18) we have

$$FX - GY = KXJ \tag{5.25}$$

$$K^T Y = F^T Y J. \tag{5.26}$$

Premultiplying (5.25) by  $Y^T$  and taking into account the transpose of (5.26), we find that  $Y^TFX$  satisfies the Lyapunov equation

$$Y^T F X = Y^T G Y + J^T Y^T F X J.$$
(5.27)

Let us show that  $F^T Y$  is invertible. Suppose that  $F^T Y$  is not invertible, so that there exists  $w \neq 0$  such that  $F^T Y w = 0$ . Then, from (5.27) we see that

$$GYw = 0 \tag{5.28}$$

$$w^T Y^T (F - G) = 0. (5.29)$$

But the stabilizability assumption implies that  $(F \ G)$  has full-row rank, so that

$$Yw = 0.$$
 (5.30)

Multiplying (5.26) on the right by w and using (5.30) we see that

$$F^T Y J w = 0. (5.31)$$

Thus, we have shown that the right null space of  $F^T Y$  is *J*-invariant. This implies that there exists an eigenvector  $w \neq 0$  of *J* in the right-null space of  $F^T Y$ , i.e.,

$$F^T Y w = 0, \quad J w = \lambda w. \tag{5.32}$$

Multiplying (5.25) on the right by w and taking into account (5.32) gives

$$(F - \lambda K)Xw = 0. \tag{5.33}$$

Since J is stable,  $|\lambda| < 1$ . The detectability assumption then implies that  $F - \lambda K$  has full-column rank, so that

$$Xw = 0. \tag{5.34}$$

Combining (5.30) and (5.34) yields

$$\binom{X}{Y}w = 0 \tag{5.35}$$

and since  $\begin{pmatrix} X \\ Y \end{pmatrix}$  has full-column rank, we must have w = 0. This is a contradiction, so that  $F^T Y$  must be invertible.

Now, if we solve for J in (5.26) and substitute it in (5.25) we obtain

$$FX = \left[G + KX(F^TY)^{-1}K^T\right]Y$$
(5.36)

from which we get

$$\begin{pmatrix} G + KX(F^TY)^{-1}K^T & F \\ F^T & 0 \end{pmatrix} \begin{pmatrix} Y \\ -X \end{pmatrix} = \begin{pmatrix} 0 \\ F^TY \end{pmatrix}.$$
 (5.37)

This implies

$$\binom{Y}{-X}(F^{T}Y)^{-1} = \binom{G + KX(F^{T}Y)^{-1}K^{T} F}{F^{T} 0} \int_{-1}^{-1} \binom{0}{I}$$
(5.38)

so that if  $P = X(F^T Y)^{-1}$ , we have

$$P = -(0 \quad I) \begin{pmatrix} G + KPK^{T}F & F \\ F^{T} & 0 \end{pmatrix}^{-1} \begin{pmatrix} 0 \\ I \end{pmatrix}$$
(5.39)

i.e., *P* satisfies the algebraic descriptor Riccati equation (4.19). Since  $Y^T F X$  solves the Lyapunov equation (5.39), it is positive semidefinite, so that

$$P = (F^{T}Y)^{-T}(Y^{T}FX)(F^{T}Y)^{-1}$$
(5.40)

is also positive semidefinite.

# VI. AN ADJUSTED ESTIMATE TO ACCOUNT FOR "FUTURE" DYNAMICS

The estimation problem we have considered in the preceding sections involved the recursive computation of estimates of x(k) based on dynamics and observations only in the past and present. As we have pointed out, descriptor dynamics allow the possibility of noncausal behavior and thus it would also seem reasonable to consider the recursive computation of estimates that incorporate future dynamics. Specifically, suppose that we now define the estimate  $\hat{x}(j)$  as the ML estimate of x(j) based on the true measurements (1.2) for  $k = 0, \dots, j - 1$ , the "measurement" (3.1) provided by the prior information about x(0), and the dynamics (1.1) for all k, as opposed to  $0 \le k \le j - 1$  as we did previously. In the usual casual case, i.e.,  $l_k = n_k = n, E = I$ , the inclusion of these "future" dynamics provide no additional information about x(j), as they provide no constraints on x(j). That is, consider

$$0 = E_{i+1}x(j+1) - A_ix(j) - u(j).$$
(6.1)

If  $E_{j+1} = I$  (or more generally, if it is surjective) then since x(j + 1) is completely unknown in the ML formulation, (6.1) provides no constraint on x(j). However, if  $E_{j+1}$  is singular, (6.1) *does* provide nontrivial information about x(j) (e.g., consider the extreme case of  $E_{j+1} = 0$ ). In general, of course, the situation is even more complex, since x(j + 1) may *also* be subject to constraints due to dynamics further into the future. In the general time-varying case there is no bound on how far into the future one must look in order to capture all possible dynamics. In such a particular case, what we would need to do each time is to filter backward the "measurements" corresponding to future dynamics in order to obtain the correct adjustment to the forward filtered estimate developed in the previous sections. However, in the time-invariant case, the structure of the information provided by the future dynamics, i.e.,

$$E_{x}(k+1) = A_{x}(k) + u(k) \qquad k \ge j \qquad (6.2)$$

is independent of j. In this section, we show that in this case we can replace the effect of future dynamics with just one observation

$$Fx(j) = w(j) \tag{6.3}$$

where w is zero mean, and F and the covariance V of w are time invariant. Thus, the problem is to find F and V in terms of E, A and Q (the covariance of u).

Rewriting (6.2) as a matrix equation we obtain

$$\begin{pmatrix} -A & E & & \\ & -A & E & \\ & & & -A & E \\ & & & & \cdot \begin{pmatrix} x(j) \\ x(j+1) \\ x(j+2) \\ \vdots \end{pmatrix} = \begin{pmatrix} u(j) \\ u(j+1) \\ u(j+2) \\ \vdots \end{pmatrix}.$$
 (6.4)

The relation (6.4) provides some information not only about x(j), but also about the vectors x(k) for k > j, which are not directly of interest and can be viewed as exogenous variables. In order to isolate the information about x(j) that is contained in (6.4), our first step will be to bring (6.4) to the form (2.39), so that Lemma 2.4 can be applied. This requires using block-row manipulations to eliminate the vectors x(k) with k > j from as many equations as we can, thereby enabling us to drop the remaining measurements. Specifically, suppose that

$$\begin{pmatrix} T_0 & T_1 & T_2 & \cdots \end{pmatrix} \begin{pmatrix} -A & E & & \\ & -A & E & \\ & & & -A & E \\ & & & & \vdots & \vdots \end{pmatrix}$$
$$= \begin{pmatrix} -T_0 A & 0 & 0 & \cdots \end{pmatrix}. \quad (6.5)$$

Then, from (6.4), we get

$$-T_0 A x(j) = \sum_{i} T_i u(j+i)$$
 (6.6)

which is of the form (6.3). So the problem becomes one of finding the highest-row rank matrix satisfying (6.5). We can rewrite (6.5) as

$$T(z)(zE - A) = -T_0A$$
 (6.7)

where  $T(z) = T_0 + zT_1 + z^2T_2 + \cdots$  and thus we need to find the polynomial matrix T(z) of largest rank such that

$$T(z)(zE - A) = \text{constant matrix.}$$
 (6.8)

Let us denote the unknown right-hand side of (6.8) by F, and let U(z) and S be, respectively, unimodular and

1338

permutation matrices for which

$$U(z)(zE - A)S = \begin{pmatrix} N(z) & K(z) \\ 0 & 0 \end{pmatrix}$$
(6.9)

where N(z) is square and invertible. Then, if we denote

$$FS = \begin{pmatrix} F_1 & F_2 \end{pmatrix} \tag{6.10a}$$

$$T(z)U^{-1}(z) = (T_1(z) \quad T_2(z))$$
 (6.10b)

we must have

$$T_1(z)N(z) = F_1$$
 (6.11a)

$$T_1(z)K(z) = F_2$$
 (6.11b)

which implies that

$$K(z)N^{-1}(z)F_1 = F_2 (6.12)$$

or equivalently

$$\begin{pmatrix} F_1 & F_2 \end{pmatrix} \begin{pmatrix} K(z)N^{-1}(z) \\ -I \end{pmatrix} = 0.$$
 (6.13)

Constant solutions  $(F_1 F_2)$  to (6.13) can be constructed by noting that if

$$\begin{pmatrix} K(z)N^{-1}(z)\\ -I \end{pmatrix} = \frac{1}{p(z)} \left(\sum_{i=1}^m L_i z^i\right) \quad (6.14)$$

where p(z) is a scalar polynomial and  $L_i$ 's are constant matrices, then (6.13) is equivalent to

$$(F_1 \quad F_2)(L_0 \quad L_1 \quad \cdots \quad L_m) = 0.$$
 (6.15)

Let  $(\tilde{F}_1 \ \tilde{F}_2)$  be a highest rank solution to (6.15). Let W be the highest rank (full-row rank) matrix for which

$$W\tilde{F}_1 N^{-1}(z) =$$
 polynomial. (6.16)

Then, let

$$F_1 = W\tilde{F}_1, \quad F_2 = W\tilde{F}_2.$$
 (6.17)

We get

$$F = \begin{pmatrix} F_1 & F_2 \end{pmatrix} S^{-1} \tag{6.18}$$

and

$$T(z) = \begin{pmatrix} F_1 N^{-1}(z) & T_2(z) \end{pmatrix} U(z)$$
(6.19)

where  $T_2(z)$  is any arbitrary polynomial matrix. It turns out that, without loss of generality, we can pick  $T_2(z) = 0$ . This is due to an implicit assumption that was made throughout this paper, namely that the dynamic equations are consistent for all possible choices of inputs u(k), i.e., no constraints on the inputs are imposed by the dynamic equations. It is straightforward to verify that this requires

$$\operatorname{Left-ker} Q \supset \operatorname{Left-ker} [zE - A] \tag{6.20}$$

which is called the *compatibility* assumption.

To see why the compatibility assumption implies that the choice of  $T_2(z)$  does not matter, simply note that thanks to (6.20) we have

$$\begin{pmatrix} 0 & T_2(z) \end{pmatrix} U(z)Q = 0.$$
 (6.21)

Finally, we get

$$Fx(j) = \sum_{i=1}^{p} T_i u(j+i)$$
 (6.22)

where  $\sum_{i=1}^{p} T_i z^i = T(z)$ . Thus, we obtain (6.3) with

$$V = \sum_{i=1}^{p} T_i Q T_i^T.$$
 (6.23)

Using (6.3), we can construct the "true" or "adjusted" descriptor Kalman estimate by correcting the result of the Kalman filter to incorporate this additional observation. In particular, using the methods developed in the previous sections, we construct the optimal estimate  $\hat{x}(j)$  of x(j) based on past dynamics and observations. This "information" is completely coded by the observation

$$\hat{x}(j) = x(j) + \nu(j)$$
 (6.24)

where  $\nu(j)$  is a zero-mean Gaussian vector with covariance  $P_j$ , where  $P_j$  satisfies the Riccati equation described previously. If we now add future dynamics, we have to find the optimal estimate of x(j) based on the observation

$$\begin{pmatrix} \hat{x}(j) \\ 0 \end{pmatrix} = \begin{pmatrix} I \\ F \end{pmatrix} x(j) + \begin{pmatrix} \nu(j) \\ w(j) \end{pmatrix}.$$
(6.25)

Since future dynamics are independent of past dynamics and observations, the new Kalman estimate  $\hat{x}(j)$  and the corresponding error covariance  $\hat{P}_i$  are given by

$$\check{x}(j) = (0 \ 0 \ I) \begin{pmatrix} P_j \ 0 \ i \\ 0 \ V \ F \\ I \ F^T \ 0 \end{pmatrix}^{\dagger} \begin{pmatrix} \hat{x}(j) \\ 0 \\ 0 \end{pmatrix} \quad (6.26)$$

and

$$\check{P}_{j} = -(0 \quad 0 \quad I) \begin{pmatrix} P_{j} & 0 & i \\ 0 & V & F \\ I & F^{T} & 0 \end{pmatrix}^{\mathsf{T}} \begin{pmatrix} 0 \\ 0 \\ I \end{pmatrix}. \quad (6.27)$$

## VII. CONCLUSIONS

In this paper, we have derived Kalman filtering recursions for a general class of discrete-time descriptor systems where the noise covariances were allowed to be singular. By using a Hamiltonian (or dual) formulation of the ML estimation problem, the optimal filter and the associated Riccati equation for the error covariance were expressed in 3-block form. In the time-invariant case, the asymptotic behavior of the optimal filter was examined and characterized in terms of the corresponding 3-block algebraic Riccati equation. Finally, under standard detectability and stabilizability conditions, it was shown that the positive semidefinite solution of the algebraic Riccati equation could be obtained by constructing the generalized Schur form of a 3-block matrix pencil.

Although we have focused primarily on descriptor systems, it is worth noting that because of the 3-block forms

1339

we have introduced, our results already present a number of advantages over existing Kalman filtering techniques for systems with standard dynamics (E = I) but with singular measurement noise. For example, in the absence of redundant perfect information, the 3-block filter and Riccati equations of Theorem 3.1 require only standard matrix inverses, whereas solutions proposed until now require the use of pseudo inverses (see [12, section VII]).

One obvious direction in which the results of our paper can be extended consists in dualizing our results by considering the descriptor LQ control problem. Preliminary results in this direction appear in [30, section VI]. Other interesting results for the descriptor LO control problem have been derived in [7] and [22]. Another possible extension would involve considering the continuous-time descriptor Kalman filtering problem. Unfortunately, the continuous-time version of the problem discussed here may not be completely meaningful. This is due to the fact that unlike the discrete-time case, where the singularity of the system dynamics gives rise to a noncausal impulse response, for continuous-time systems the singularity manifests itself by the fact that the output contains derivatives of the system input. White Gaussian noise is the input for the filtering problem, the output will contain white-noise derivatives, thereby necessitating a formulation of the filtering problem in terms of generalized stochastic processes.

#### APPENDIX A

## SOME RESULTS ON BLOCK PSEUDO INVERSES

Here we summarized and specialize several of the results in [11] concerning the generalized inverse [in the sense of (2.20)] of the matrix

$$Z = \begin{pmatrix} R & H \\ H^T & 0 \end{pmatrix}$$
(A.1)

when H has full-column rank, i.e., when  $(H^TH)$  is invertible. Let

$$Z^{\dagger} = \begin{pmatrix} W & U \\ U^T & T \end{pmatrix}$$
(A.2)

denote any symmetric matrix satisfying (2.20), which in this case reduces to

$$RWR + RUH^T + HU^TR + HTH^T = R$$
 (A.3)

$$RWH + HU^TH = H \quad (A.4)$$

$$H^T W H = 0. \quad (A.5)$$

In [11] the following results are proved

$$RWH = 0 \tag{A.6}$$

$$D = R + RWR$$
 is uniquely determined by R and H

$$HTH^T = -D \tag{A.8}$$

$$RUH^{T} = D. \tag{A.9}$$

Next, note that from (A.8) and the invertibility of  $H^T H$ , we have that T is also unique and given by

$$T = -(H^{T}H)^{-1}H^{T}DH(H^{T}H)^{-1}.$$
 (A.10)

Also, from (A.4), (A.6) and the invertibility of  $H^{T}H$ 

$$U^{T}H = (H^{T}H)^{-1}(H^{T}H) = I$$
 (A.11)

proving (4.27) of Lemma 4.2, so that  $U^T$  is a left inverse<sup>7</sup> of *H*. Thus, from (2.21)

$$\mathscr{M}(\hat{x}_{\mathsf{ML}}) = (0 \quad I) \begin{pmatrix} W & U \\ U^T & T \end{pmatrix} \begin{pmatrix} H \\ 0 \end{pmatrix} x = x.$$
(A.12)

Also from (2.21)

$$P_{\rm ML} = (0 \quad I) Z^{\dagger} \begin{pmatrix} R & 0 \\ 0 & 0 \end{pmatrix} Z^{\dagger} \begin{pmatrix} 0 \\ I \end{pmatrix}.$$
(A.13)

However, from (A.8), (A.9), and (A.11) we see that

$$P_{\rm ML} = U^T D H (H^T H)^{-1} = -U^T H T H^T H (H^T H)^{-1} = -T \quad (A.14)$$

. 1

proving (2.22) as well as (4.26) of Lemma 4.2.

Finally, to prove identity (2.25), we note that by summing (A.8) and (A.9) and postmultiplying by  $H(H^TH)^{-1}$ , we obtain

$$RU + HT = 0 \tag{A.15}$$

which gives (2.25) when combined with (A.11).

## APPENDIX B DERIVATION OF (4.45)

Let

Then

$$\Omega_{i} = \begin{pmatrix} AP_{i}A^{T} + Q & -S & E \\ -S^{T} & R & C \\ E^{T} & C^{T} & 0 \end{pmatrix}, \quad i = 1, 2. \quad (B.1)$$

$$P^{1} - P^{2} = (0 \quad 0 \quad I)(\Omega_{2}^{\dagger} - \Omega_{1}^{\dagger}) \begin{pmatrix} 0 \\ 0 \\ I \end{pmatrix}. \quad (B.2)$$

From identity (2.25), there exists  $\Delta$  such that

$$\begin{pmatrix} 0\\0\\I \end{pmatrix} = \Omega_2 \Delta \tag{B.3}$$

101

so that

$$\Omega_2 \Omega_2^{\dagger} \begin{pmatrix} 0\\0\\I \end{pmatrix} = \Omega_2 \Omega_2^{\dagger} \Omega_2 \Delta \qquad (B.4)$$

which using the property (2.20) of pseudo inverses yields

$$\Omega_2 \Omega_2^* \begin{pmatrix} 0\\0\\I \end{pmatrix} = \begin{pmatrix} 0\\0\\I \end{pmatrix}. \tag{B.5}$$

<sup>7</sup> Note that  $U^T$  is not an *arbitrary* left inverse of H as there is the additional constraint (A.9) that U must satisfy.

Similarly we can show that

$$(0 \ 0 \ I)\Omega_1\Omega_1^{\dagger} = (0 \ 0 \ I).$$
 (B.6)

Now, by using identities (B.5) and (B.6) in (B.2) we get

$$P^{1} - P^{2} = (0 \quad 0 \quad I)\Omega_{2}^{\dagger}(\Omega_{1} - \Omega_{2})\Omega_{1}^{\dagger}\begin{pmatrix}0\\0\\I\end{pmatrix}. (B.7)$$

But

$$\Omega_{1} - \Omega_{2} = \begin{pmatrix} A(P^{1} - P^{2})A^{T} & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix}$$
$$= \begin{pmatrix} I \\ 0 \\ 0 \end{pmatrix} A(P^{1} - P^{2})A^{T}(I - 0 - 0). \quad (B.8)$$

Thus

$$P^{1} - P^{2} = \left( \begin{pmatrix} 0 & 0 & I \end{pmatrix} \Omega_{2}^{\dagger} \begin{pmatrix} I \\ 0 \\ 0 \end{pmatrix} \right) \mathcal{A}(P^{1} - P^{2}) \mathcal{A}^{T}$$
$$\cdot \left( \begin{pmatrix} I & 0 & 0 \end{pmatrix} \Omega_{1}^{\dagger} \begin{pmatrix} 0 \\ 0 \\ I \end{pmatrix} \right)^{T}$$

$$= (L^{2}A)(P^{1} - P^{2})(L^{1}A^{T}).$$
 (B.9)

## REFERENCES

- [1] M. B. Adams, A. S. Willsky, and B. C. Levy, "Linear estimation of boundary value stochastic processes—Part I: The role and con-struction of complementary models," *IEEE Trans. Automat. Contr.*,
- vol. AC-29, pp. 803–810, Sept. 1984. [2] B. D. O. Anderson and J. B. Moore, *Optimal Filtering*. Englewood Cliffs, NJ: Prentice-Hall, 1979.
- W. F. Arnold and A. J. Laub, "Generalized eigenproblem algo-[3] rithms and software for algebraic Riccati equations," Proc. IEEE, vol. 72, pp. 1746-1754, Dec. 1984.
- Voi. 72, pp. 1740-1754, Dec. 1964.
  D. J. Bender and A. J. Laub, "The linear-quadratic optimal regulator for descriptor systems," *IEEE Trans. Automat. Contr.*, vol. AC-32, pp. 672-688, Aug. 1987.
  —, "The linear-quadratic optimal regulator for descriptor systems: Discrete-time case," *Automatica*, vol. 23, pp. 71-85, Jan. 1967. [4]
- [5] 1987.
- E. Bryson, Jr. and Y.-C. Ho, Applied Optimal Control. [6] Waltham, MA: Ginn and Co., 1969.
- P. Bernhard, J. Grimm, and X.-M. Wang, "Commande optimale [7] linéaire quadratique des systèmes implicites," APII, vol. 24, pp. 17-34, 1990
- P. Bernhard and X.-M. Wang, "Filtrage des systèmes implicites [8] linéaires discrets," C. R. Acad. Sc., Paris, t. 304, série I, pp. 351-354, 1987.
- S. L. Campbell, "Singular Systems of Differential Equations," [9] (Research Notes in Math., No. 40). San Francisco, CA: Pitman, 1980.
- "Singular of Differential Equations II," (Research Notes in [10] Math., No. 61). San Francisco, CA: Pitman, 1982.
- S. L. Campbell and C. D. Meyer, Generalized Inverses of Linear [11] Transformations. London: Pitman, 1979.
- D. E. Catlin, Estimation, Control, and the Discrete Kalman Filter. New York: Springer-Verlag, 1989. [12]
- L. Dai, "Filtering and LQG problems for discrete-time stochastic singular systems," *IEEE Trans. Automat. Contr.*, vol. 34, pp. [13] 1105-1108, Oct. 1989.
- [14] A. Emami-Naeini and G. F. Franklin, "Deadbeat control and tracking of discrete-time systems," *IEEE Trans. Automat. Contr.*, vol. AC-27, pp. 176–181, Feb. 1982.
- G. H. Golub and C. F. Van Loan, Matrix Computations, 2nd ed., [15] Baltimore, MD: The Johns Hopkins Univ. Press, 1989.

- [16] A. J. Krener, "A causal realization theory, Part I: Linear deterministic systems," SIAM J. Contr. Optimiz., vol. 25, pp. 499-525, May 1987
- V. Kucera, "The discrete Riccati equation of optimal control," [17] Kybernetika, vol. 8, 1972.
- [18] F. L. Lewis and V. G. Mertzios, Eds., "Recent advances in singular systems," Special issue, Circuits, Syst., Signal Proc., vol. 8, 1989
- [19] D. G. Luenberger, "Dynamic systems in descriptor form," IEEE Trans. Automat. Contr., vol. AC-22, pp. 312-321, June 1977.
- [20] "Time-invariant descriptor systems," Automatica, vol. 14, pp. 473-480, Sept. 1978.
- [21] -, "Boundary recursion for descriptor variable systems," IEEE Trans. Automat. Contr., vol. 34, pp. 287-292, Mar. 1989.
- [22] V. Mehrmann, "Existence, uniqueness, and stability of solutions to singular linear quadratic optimal control problems," Linear Algebra Appl., vol. 121, pp. 291-331, 1989.
- [23] R. Nikoukhah, "A deterministic and stochastic theory for two-point boundary-value descriptor systems," Ph.D. dissertation, Dept. Elec. Eng. Comp. Sci., Mass. Inst. Technol., Cambridge, MA, Sept. 1988.
- [24] R. Nikoukhah, M. B. Adams, A. S. Willsky, and B. C. Levy, "Estimation for boundary-value descriptor systems," Circuits, Syst., Signal Proc., vol. 8, pp. 25-48, 1989.
- R. Nikoukhah, B. C. Levy, and A. S. Willsky, "Stability, stochastic [25] stationarity, and generalized Lyapunov equations for two-point boundary-value descriptor systems," IEEE Trans. Automat. Contr., vol. 34, pp. 1141-1152, Nov. 1989.
- [26] "Smoothing algorithms for boundary-value descriptor sys-
- [27] ability, and observability," Int. J. Contr., vol. 46, pp. 1715-1737, Nov. 1987.
- [28] , "Generalized Riccati equations for two-point boundary-value descriptor systems," in Proc. 26th IEEE Conf. Decision Contr., Los Angeles, CA, Dec. 1987, pp. 1140-1141.
- ----, "Reachability, observability and minimality for shift-in-variant two-point boundary-value descriptor systems," Circuits, [29] Syst., Signal Proc., vol. 8, pp. 313-340, 1989.
- "Kalman filtering and Riccati equations for descriptor sys-[30] tems," Tech. Rep. 1186, Institut National de Recherche en Informatique et Automatique, Rocquencourt, France, Mar. 1990.
- [31] T. Pappas, A. J. Laub, and N. R. Sandell, Jr., "On the numerical solution of the discrete-time algebraic Riccati equation," IEEE Trans. Automat. Contr., vol. 25, pp. 631–641, Aug. 1980.
- P. Van Dooren, "A generalized eigenvalue approach for solving Riccati equations," *SIAM J. Sci. Stat. Comput.*, vol. 2, pp. 121–135, [32] 1981
- X-M. Wang and P. Bernhard, "Filtrage et lissage des systèmes [33] discrets," Tech. Rep. 1083, Institut National de Recherche en Informatique et Automatique, Rocquencourt, France, Aug. 1989.
- P. Whittle, Prediction and Regulation by Linear Least-Square Meth-[34] ods, 2nd ed. Minneapolis, MN: University of Minnesota Press, 1983
- J. C. Willems, "A framework for the study of dynamical systems," [35] in Realization and Modelling in System Theory, vol. 1, M. A. Kaashoek, J. H. Van Schuppen, and A. C. M. Ran, Eds., Boston, MA: Birkhauser-Verlag, pp. 43-60, 1990, also in Proc. MTNS-89.



Ramine Nikoukhah (S'87-M'90) was born in Teheran, Iran, on March 30, 1961. He received the B.S. degree from the University of California, Davis, in 1982, the degree of Maitrise from the University of Paris-Sud, Orsay, France, in 1983, the M.S. and Engineer's degrees in 1986, and the Ph.D. degree, in 1988, from the Massachusetts Institute of Technology, Cambridge, all in electrical engineering and computer science.

During 1983 he was a Research Engineer at

TAU Co., Los Gatos, CA, where he worked in the area of flight control. Since 1988, he has been with the Institut National de Recherche en Informatique et en Automatique (INRIA), Rocquencourt, France. He has also consulted for Matra Co. and has held teaching positions at various engineering schools in France. His research interests are in the areas of control, signal processing, and algebraic system theory.

#### 1342

#### IEEE TRANSACTIONS ON AUTOMATIC CONTROL, VOL. 37, NO. 9, SEPTEMBER 1992



Alan S. Willsky (S'70–M'73–SM'82–F'86) received the S.B. degree and the Ph.D. degree from the Massachusetts Institute of Technology Cambridge, in 1969 and 1973, respectively. He joined the faculty at M.I.T., in 1973 and his present position is Professor of Electrical Engineering. From 1974 to 1981, he served as Assistant Director of the M.I.T. Laboratory for Information and Decision Systems. He is also a founder and member of the Board of Directors of Alphatech, Inc., Burlington, MA. He has held

visiting positions at Imperial College, London, England, L'Université de Paris-Sud, France, and the Institut de Recherche en Informatique et Systèmes Aléatoires in Rennes, France. His present research interests are in problems involving multidimensional and multiresolution estimation and imaging, discrete-event systems, and the asymptotic analysis of control and estimation systems. He was Program Chairman for the 17th IEEE Conference on Decision and Control, has been an Associate Editor of several journals including the IEEE TRANSACTIONS ON AUTO-MATIC CONTROL, has served as a member of the Board of Governors and Vice President for Technical Affairs of the IEEE Control Systems Society, was Program Chairman for the 1981 Bilateral Seminar on Control Systems held in the People's Republic of China, and was Special Guest Editor of the 1992 special issue of the IEEE TRANSACTIONS ON INFORMATION THEORY on wavelet transforms and multiresolution signal analysis. Also in 1988, he was made a Distinguished Member of the IEEE Control Systems Society. In addition, he has given several plenary lectures at major scientific meetings including the 20th IEEE Conference on Decision and Control and the 1991 IEEE International Conference on Systems Engineering.

Dr. Willsky is the author of the research monograph, Digital Signal Processing and Control and Estimation Theory and is coauthor of the undergraduate text Signals and Systems. In 1975, he received the Donald P. Eckman Award from the American Automatic Control Council. He was awarded the 1979 Alfred Noble Prize by the ASCE and the 1980 Browder J. Thompson Memorial Prize Award by the IEEE for a paper excerpted from his monograph.



Bernard C. Levy (S'74-S'78-M'78-SM'90) was born in Princeton, NJ, on July 31, 1951. He received the diploma of Ingénieur Civil des Mines from the Ecole Nationale Supérieure des Mines, Paris, France, and the Ph.D. degree in electrical engineering from Stanford University, Stanford, CA.

While at Stanford University, he held an IN-RIA Fellowship and also worked as a Teaching Assistant, Research Assistant, and Instructor. From June 1979 to June 1987, he was an Assis-

tant, and then Associate Professor in the Department of Electrical Engineering and Computer Science at M.I.T., Cambridge. Since July 1987, he has been with the University of California, Davis, where he is currently a Professor of Electrical Engineering. His research interests are in the areas of multidimensional and statistical signal processing, inverse problems estimation, detection, and scientific computing.

inverse problems, estimation, detection, and scientific computing. Dr. Levy is an Associate Editor of the journal, *Multidimensional* Systems and Signal Processing.